Synchronization of Hyperchaotic Memristor-Based Chua's Circuits

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Abstract This paper further investigates the problem of synchronization of hyperchaotic memristor-based Chua's circuits. An active control method is employed to design a controller to achieve the global synchronization of two identical memristor-based systems. Based on Lyapunov stability theory, a sufficient condition is given to guarantee the stability of the synchronization error system.

Keywords Memristor-based systems • Synchronization • Chua's circuit

1 Introduction

The existence of the memristor as the fourth fundamental circuit element included along with the resistor, capacitor, and inductor was predicated by Chua in 1971 [1]. Until 2008, the Hewlett-Packard (HP) research team announced that they had realized a prototype of memristor-based on nanotechnology [2]. Many researchers focus on the memristor because of its potential applications in programmable logic, signal processing, neural networks, control systems, reconfigurable computing, Brain-computer interfaces, and RFID [3–9].

Recently, the research on circuits based on memristor is becoming a focal topic [10–19]. Itoh and Chua presented a fourth-order memristor-based Chua's oscillator by replacing Chua's diode with an active two-terminal circuit consisting of a conductance and a flux-controlled memristor [10]. Pershin and Di Ventra introduced an approach to use memristors in programmable analog circuits [11]. Rak and

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Cserey presented a new simulation program with integrated circuit emphasis macromodel of the recently physically implemented memristor [12]. Petras presented fractional-order memristor-based Chua's circuit [13]. The hyperchaotic behavior in memristor-based Chua's circuit is performed with the help of nonlinear tools [19]. In this letter, we will study the problem of synchronization of memristor-based Chua's systems. Based on the feedback control method, we design a controller to guarantee the exponentially stability of the synchronization error system.

The rest of the paper is organized as follows. In Sect. 2, a memristor-based system is introduced. And using feedback control method, a general convergence criterion for stabilization of synchronization error system is established. Conclusions are finally drawn in Sect. 3.

2 Problem Formulation and Preliminaries

Referring to [19], Andrew L. Fitch proposed a circuit by adding an inductor in parallel with conductance—G and fourth-order memristor-based canonical oscillato. The equations for the circuit are described by

$$\begin{cases} \frac{dq_{l_2}(t)}{dt} = \frac{1}{L_2}\phi_{c2}(t), \\ \frac{d\phi_{c2}(t)}{dt} = \frac{1}{C_2}\left(-q_{l_2}(t) - G\phi_{c2}(t) - q_{l_1}(t)\right), \\ \frac{dq_{l_1}(t)}{dt} = \frac{1}{L_1}\left(\phi_{c2}(t) - q_{l_2}(t)R - \phi_{c1}(t)\right), \\ \frac{d\phi_{c1}(t)}{dt} = \frac{1}{C_1}\left(-q_{l_1}(t)R - a\phi_{c1}(t) - b\phi_{c1}^{-3}(t)\right). \end{cases}$$
(1)

The system Eq. (1) can be reorganized by

$$\begin{cases} \frac{d\phi_{c1}(t)}{dt} = -\tau v_1(t), \\ \frac{dv_1(t)}{dt} = \frac{1}{C_1} (i_{L1}(t) - W(\phi_{c1}(t)) v_1(t)), \\ \frac{dv_2(t)}{dt} = \frac{1}{C_2} (Gv_2(t) - i_{L1}(t) - i_{L2}(t)), \\ \frac{di_{L1}(t)}{dt} = \frac{1}{L_1} (v_2(t) - v_1(t) - Ri_{L1}(t)) \\ \frac{di_{L2}(t)}{dt} = \frac{1}{L_2} v_2(t). \end{cases}$$

$$(2)$$

where τ is an integration constant which introduced to rescale the values of voltage into practical range, and $W(\phi_{c1}(t)) = a + 3b\phi_{c1}^2(t)$.

Letting $x_1(t) = \phi_{c1}(t)$, $x_2(t) = v_1(t)$, $x_3(t) = v_2(t)$, $x_4(t) = i_{L1}(t)$, $x_5(t) = i_{L2}(t)$, system (2) can be further rewritten as

$$\begin{cases} \dot{x}_{1}(t) = -\tau x_{2}(t), \\ \dot{x}_{2}(t) = \frac{1}{C_{1}} \left(x_{4}(t) - a x_{2}(t) - 3b x_{1}^{2}(t) x_{2}(t) \right), \\ \dot{x}_{3}(t) = \frac{1}{C_{2}} \left(G x_{3}(t) - x_{4}(t) - x_{5}(t) \right), \\ \dot{x}_{4}(t) = \frac{1}{L_{1}} \left(x_{2}(t) - x_{3}(t) - R x_{4}(t) \right), \\ \dot{x}_{5}(t) = \frac{1}{L_{2}} x_{3}(t). \end{cases}$$
(3)

When $L_1 = 10mH$, $L_2 = 60mH$, $C_1 = 6.8nF$, $C_2 = 15nF$, G = 0.0005S, a = 0.00067, b = 0.000029, $R = 65\Omega$, $\tau = 26,000$, existence of hyperchaos.

The system (3) can be rewritten with linear part and nonlinear part as follows:

$$\dot{x}(t) = Ax(t) + f(x(t)),$$
 (4)

where $x(t) = (x_1(t), x_2(t), x_3(t), x_4(t), x_5(t))^T$,

$$A = \begin{bmatrix} 0 - \tau & 0 & 0 & 0 \\ 0 & \frac{-a}{C_1} & 0 & \frac{1}{C_1} & 0 \\ 0 & 0 & \frac{G}{C_1} & \frac{-1}{C_2} & \frac{-1}{C_2} \\ 0 & \frac{1}{L_1} & \frac{-1}{L_1} & \frac{-R}{L_1} & 0 \\ 0 & 0 & \frac{-1}{L_2} & 0 & 0 \end{bmatrix}, \quad A = \begin{pmatrix} 0 & -\tau & 0 & 0 & 0 \\ 0 & -a/C_1 & 0 & 1/C_1 & 0 \\ 0 & 0 & G/C_2 & -1/C_2 & -1/C_2 \\ 0 & 1/L_1 & -1/L_1 & -R/L_1 & 0 \\ 0 & 0 & 1/L_2 & 0 & 0 \end{pmatrix}$$

and

$$f(x(t)) = \begin{pmatrix} 0 \\ -3bx_1^2 x_2/C_1 \\ 0 \\ 0 \\ 0 \end{pmatrix}.$$

As for vector function f(x), assume that for any $x, y \in \Omega$ we have

$$|f_i(x) - f_i(y)| \le L_{\max} |x - y|, \ i = 1, \ 2, \ 3, \ 4$$
(5)

The above condition is considered as the uniform Lipszchitz condition, and $L_{\text{max}} > 0$ refers to the uniform Lipschitz constant.

We construct the response system as below:

$$\dot{y}(t) = Ay(t) + f(y(t)) + u(t)$$
 (6)

where $y(t) = (y_1(t), y_2(t), y_3(t), y_4(t))^T$ is the response state, u(t) is the control gain defined by:

$$u(t) = k(y(t) - x(t)).$$

where $\tau > 0$ is the propagation delay, *k* denotes control strength. Let e(t) = y(t) - x(t) be the synchronization error between the systems (2) and (3), then yields the error system

$$\dot{e}(t) = \dot{y}(t) - \dot{x}(t) = Ae(t) + f(y) - f(x(t)) + u(t) = Ae(t) + f(y) - f(x(t)) + ke(t)$$
(7)

We now state our main results.

Theorem 1. Suppose that there exist positive constants s_1 , g_1 such that $A + A^T + 2kI + s_1I + s_1^{-1}L_{\max}^2 + g_1I \le 0$,

Then, the synchronization error system (7) *is globally exponentially stable, and the systems* (4) *and* (6) *are globally exponentially synchronized.*

Proof. Choose the Lyapunov function as follows

$$V(t) = e(t)^{T} e(t).$$
(8)

Then the differentiation of V along trajectories of (7) is

$$\begin{split} \dot{V}(t) &= e(t)^{T} \dot{e}(t) + \dot{e}(t)^{T} e(t) \\ &= e(t)^{T} \left[A e(t) + f(y(t)) - f(x(t)) + k e(t) \right]^{T} e(t) \\ &= e(t)^{T} \left[A + A^{T} + 2kI \right] e(t) + k e(t) \right]^{T} e(t) \\ &\leq e(t)^{T} \left[A + A^{T} + 2kI \right] e(t) + s_{1} e(t)^{T} e(t) \\ &+ s_{1}^{-1} [f(y(t)) - f(x(t))]^{T} \left[f(y(t)) - f(x(t)) \right] \\ &\leq e(t)^{T} \left[A + A^{T} + 2kI + s_{1}I \right] e(t) + s_{1}^{-1} L_{\max}^{2} e(t)^{T} e(t) \\ &= e(t)^{T} \left[A + A^{T} + 2kI + s_{1}I + s_{1}^{-1} L_{\max}^{2} + g_{1}I \right] e(t) - g_{1} e(t)^{T} e(t) \\ &\leq -g_{1} e(t)^{T} e(t) \\ &= -g_{1} V(t) \end{split}$$

According to Lyapunov theory, the inequality $\dot{V}(t) \leq -g_1 V(t)$ indicate V(t) converges to zero exponentially. Furthermore, we can conclude that the synchronization error systems e(t) converges to zero globally and exponentially with a rate g_1 , and the synchronization between with system (4) and system (6) can be obtained. This completes the proof.

3 Conclusions

In this paper, the synchronization problem of memristor-based chaotic system has been discussed. A feedback controller was designed to stabilize the synchronization error system globally exponentially.

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